

*seconda serie*

*Pubblicazioni del*

# CENTRO STUDI CALCOLATRICI ELETTRONICHE

*del C. N. R.*

*presso l'Università degli studi di Pisa*

n. 118

Session 36: ACTIVE AND PASSIVE NETWORKS  
 Paper 36.3  
 19CP67-1182

A TRANSISTORIZED DIRECTIONAL COUPLER

F.Denoth  
 Centro Studi Calc.Elettr.-CNR  
 Pisa - Italy

B.Pellegrini  
 Centro Studi Elett. e Telecom.-CNR  
 Facoltà Ingegneria - Pisa - Italy

In this paper is described a directional coupler in which the analogical reconstruction of the backward wave is obtained by means of a transistor. The use of an active element has allowed one to obtain at the same time a broad band, a high directivity and a tight coupling.

The use of such a coupler in a bidirectional broad band amplifier is also proposed.

Active Directional Couplers (^)

In a transmission line the amplitude of the backward wave voltage ( $V^-$ ) is related, at any point, to the total current ( $I$ ) and voltage ( $V$ ) and to the characteristic impedance ( $Z$ ) of the line by the following relationship:

$$V^- = (V - ZI)/2 \quad (1)$$

According to eq.(1) the reconstruction of  $V^-$  can be made by picking-up both  $V$  and  $I$  and by performing operations appearing in eq.(1) in an analogic way. For the analogical reconstruction we propose the use of a junction transistor as shown in the schematic diagram of Fig.1. When  $Z_E/(1-\alpha) \gg |Z_I|$  and  $|Z_E| \gg 1/g_m$ , the circuit of Fig.1 gives an output voltage proportional to  $V^-$  if:

$$Z_V Z_I / Z_E = Z \quad (2)$$

In particular, the output voltage is equal to  $V^-$  if

$$Z_L = Z + Z_V \quad (3)$$

If we denote by  $\Delta Z_I, \Delta Z_V, \Delta Z_E$  and  $\Delta Z$  the amount by which  $Z_I, Z_V, Z_E$  and  $Z$  differ from a set of values that satisfies eq.(2), we obtain the following expression for the directivity at low frequency ( $D_o$ ):

$$D_o = 2 / (\Delta Z_E / Z_E + \Delta Z / Z - \Delta Z_I / Z_I - \Delta Z_V / Z_V) \quad (4)$$

The expressions of the coupling factor ( $C$ ) and of the directivity ( $D$ ) versus frequency are given by:

$$C = 1 / (1 + j\omega / \omega_c) \quad (5)$$

$$D = D_o / [1 - D_o (\omega / \omega_d)^2] \quad (6)$$

where  $\omega_c$  and  $\omega_d$  are cut off frequencies, functions of the circuit parameters (°).

Bidirectional Amplifiers

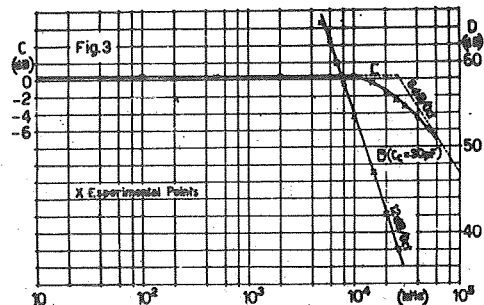
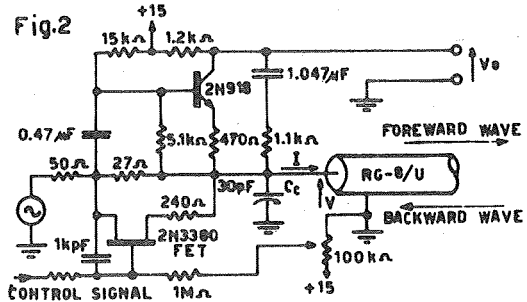
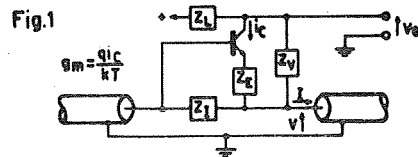
The described directional coupler can be used in a bidirectional amplifier in a conventional way. In this use it is very important to maintain a high directivity against variations of circuit parameters. This can be achieved by

supplying the system with a control signal (°). A phase detector at the output of the directional coupler gives a signal by which the value of one of the  $Z$ 's appearing in eq.(4) can be controlled. We propose the use of a FET connected at the  $Z_I$ -terminals, as appears in Fig.2.

Experimental Results

An active directional coupler has been realized according to the electric diagram of Fig.2. A theoretic analysis of the circuit (°) shows that the performances can be improved by using a capacitor of proper value connected as  $C_c$  in Fig.2.

In Fig.3 are plotted  $C$  and  $D$  versus frequency. It can be seen that the experimental measurements agree quite well with the theoretic curves.



(^) Pending patent by B.Pellegrini  
 (°) Internal Report of CSCE and CSET.