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Corresponding Author: Ms. Yuan Chen,

Corresponding Author's Institution:

First Author: Yuan Chen

Order of Authors: Yuan Chen; Hing Cheung So; Ercan Engin Kuruoglu ; Xiaolong Yang

Abstract: Parameter estimation from noisy complex-valued measurements is a significant topic in various areas of science and engineering. In this aspect, an important goal is finding an unbiased estimator with minimum variance. Therefore, variance analysis of an estimator is desirable and of practical interest. In this paper, we concentrate on analyzing the complex-valued l_p -norm minimizer with $p \geq 1$. Variance formulas for the resultant nonlinear estimators in the presence of three representative bivariate noise distributions, namely, α -stable, Student's t and mixture of generalized Gaussian models, are derived. To guarantee attaining the minimum variance for each noise process, optimum selection of p is studied. All our results are confirmed by simulations and are compared with the Cramér-Rao lower bound.

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Subject:

**“Variance Analysis of Unbiased Complex-valued Lp-Norm
Minimizer”**

by Yuan Chen, Hing Cheung So, Ercan Engin Kuruoglu and Xiao Long Yang

Dear Editor:

We would like to submit the above-mentioned paper to *Signal Processing* for consideration of publication.

We would be pleased to provide any further clarifications needed.

Yours faithfully,

Ms Yuan Chen
Prof. Hing Cheung So
Dr. Ercan Engin Kuruoglu
Prof. Xiao Long Yang

Highlights:

- The variance formulas for the complex-valued ℓ_p -norm minimization with $p \geq 1$ are derived;
- Three representative bivariate noise models, namely, symmetric α -stable, Student's t and mixture of generalized Gaussian processes, are investigated.
- To study the minimum variance of the least ℓ_p -norm estimator, the optimum p can be obtained by finding the root of a third-order polynomial;

Variance Analysis of Unbiased Complex-valued ℓ_p -Norm Minimizer

Yuan Chen¹, Hing Cheung So², *EURASIP Member*, Ercan Engin Kuruoglu³, *EURASIP Member*, Xiao Long Yang¹

Abstract

Parameter estimation from noisy complex-valued measurements is a significant topic in various areas of science and engineering. In this aspect, an important goal is finding an unbiased estimator with minimum variance. Therefore, variance analysis of an estimator is desirable and of practical interest. In this paper, we concentrate on analyzing the complex-valued ℓ_p -norm minimizer with $p \geq 1$. Variance formulas for the resultant nonlinear estimators in the presence of three representative bivariate noise distributions, namely, α -stable, Student's t and mixture of generalized Gaussian models, are derived. To guarantee attaining the minimum variance for each noise process, optimum selection of p is studied. All our results are confirmed by simulations and are compared with the Cramér-Rao lower bound.

Keywords: Variance analysis, complex-valued signals, fractional lower-order moment, ℓ_p -norm minimization, digamma function, Taylor series expansion

1. Introduction

In many areas of science and engineering, such as wireless communications, sensor array signal processing and biomedical sciences [1]–[3], observations are more conveniently modeled as complex-valued data, which have a simpler analytical form and is easier to deal with than the real-valued model. Parameter estimation for the complex-valued observations [4]–[5] is an important research topic and has attracted a great deal of attention. For numerous estimators developed in the literature, the goal is to find one which on the average yields the true value and the mean square error (MSE) between the estimate and true value is the smallest. MSE analysis of unbiased estimators is significant to help searching for one with minimum variance. Since the definition of variance relates to expectation, calculating by excessive simulations may be nonconclusive and unrealistic. To obtain the variance elegantly and correctly, approaches such as Taylor series expansion (TSE) on the estimates [6] and on the error function [7] are proposed, which are verified in [8]. However, they only consider estimators in the presence of complex Gaussian noise, which cannot describe other types of disturbances, especially those with impulsive nature, appeared in many fields [9]–[11]. For example, symmetric α -stable (SaS), Student's t and mixture of generalized Gaussian (MGG) processes are commonly employed to model the complex-valued impulsive noise.

Due to the high computational complexity and difficult implementation in case of lacking analytical prob-

¹School of Computer & Communication Engineering, University of Science & Technology Beijing, Beijing, China

²Department of Electronic Engineering, City University of Hong Kong, Hong Kong SAR, China

³Italian National Council of Research (ISTI-CNR), Pisa, Italy

ability density function (PDF), the maximum likelihood (ML) estimator may not be the proper choice in the presence of non-Gaussian noise models. Alternatively, a robust estimation methodology, namely, M -estimator is proposed by Huber [12], which generalizes the ML estimator by replacing the logarithm of the likelihood function by an arbitrary ρ -function. Defining the ρ -function as the ℓ_p -norm of residual [13], the least ℓ_p -norm estimator with $p \geq 1$ is widely utilized in the environment with non-Gaussian noise.

In this paper, we investigate the performance of the ℓ_p -norm minimizer using the complex-valued observations. The S α S, Student's t and MGG noise models are considered. To guarantee the unbiasedness of the least ℓ_p -norm estimator, we assume all discussed models are symmetric with zero location parameter [14]. For the ℓ_p -norm minimizer, since the variance formulas should be a function of p , the selection of optimum p is required to ensure achieving the minimum variance criterion. In [15], the relationship between kurtosis and optimum p is investigated, however, the results cannot be used when the kurtosis does not exist such as in the presence of S α S noise. The optimal selection of p for S α S distribution is discussed in [16], but optimum estimation performance for other noise models is not addressed. Furthermore, although [14] shows that the optimum p can be obtained by a polynomial equation with a very large degree of nonlinearity. it suffers a high computational cost in solving such a high-order polynomial function. To overcome the complexity problem, in this work we simplify the optimal choice problem into finding root of a low-order function using the property of gamma function. Although we focus on the the complex-valued estimation problem in this paper, our result can be extended to the real-valued case.

The rest of this paper is organized as follows. In Section 2, we briefly review the bivariate S α S, Student's t and MGG models. The variance formulas of the least ℓ_p -norm estimator are devised in Section 3. The selection of p for different noise processes with the minimum variance is also examined. In Section 4, simulations are provided to validate the derived variance formulas with comparison to the Cramér-Rao lower bound (CRLB). Finally, conclusions are drawn in Section 5.

2. Review of Well-known Distributions

According to the analysis in [14], the variance formula needs the fractional lower-order moment (FLOM) of the noise. Since a complex random variable corresponds to a bivariate distribution, in this section, we review the bivariate S α S, Student's t and MGG distributions.

2.1. S α S Distribution

Let $q = \Re\{q\} + j\Im\{q\}$ follow a bivariate S α S distribution. Its characteristic function has the form of [17]:

$$\varphi(t) = \exp(j(\delta_r \Re\{t\} + \delta_i \Im\{t\}) - \gamma(\Re\{t\}^2 + \Im\{t\}^2)^{\frac{\alpha}{2}}), \quad (1)$$

where $\alpha \in (1, 2]$ is the stability parameter, controlling the impulsiveness of the distribution, $(\delta_r, \delta_i) \in (-\infty, \infty) \times (-\infty, \infty)$ denote the location parameters, and $\gamma > 0$ is the dispersion parameter which determines the spread of the distribution. Note that we set $\delta_r = \delta_i = 0$ here because the location parameter of noise term is assumed zero.

The FLOMs of real and imaginary parts of q , denoted by $E\{|q|^{l-2}\Re\{q\}^2\}$ and $E\{|q|^{l-2}\Im\{q\}^2\}$, are [18]:

$$E\{|q|^{l-2}\Re\{q\}^2\} = D_\alpha(l, \alpha)\gamma^{\frac{l}{\alpha}}, \quad (2)$$

$$E\{|q|^{l-2}\Im\{q\}^2\} = D_\alpha(l, \alpha)\gamma^{\frac{l}{\alpha}}, \quad -2 < l < \alpha, \quad (3)$$

where $E\{\cdot\}$ denotes the expectation operator and

$$D_\alpha(l, \alpha) = \frac{\Gamma(\frac{l+2}{2})\Gamma(1 - \frac{l}{\alpha})}{\Gamma(1 - \frac{l}{2})} 2^{l-1}, \quad (4)$$

with $\Gamma(\cdot)$ being the gamma function.

2.2. Student's t Distribution

For a zero-mean complex-valued variable $q = \Re\{q\} + j\Im\{q\}$, which follows the bivariate Student's t distribution, the PDF is [10]:

$$f(q) = \frac{1}{2\pi\eta^2} \left(1 + \frac{1}{\nu} \left(\frac{\Re\{q\}^2 + \Im\{q\}^2}{\eta^2} \right) \right)^{-\frac{\nu+2}{2}}, \quad (5)$$

where $\nu > 0$ is the number of degrees of freedom, and $\eta > 0$ denotes the scaling parameter determining the spread of the PDF.

The FLOMs of $\Re\{q\}$ and $\Im\{q\}$ are:

$$E\{|q|^{l-2}\Re\{q\}^2\} = D_t(l, \nu)\eta^l, \quad (6)$$

$$E\{|q|^{l-2}\Im\{q\}^2\} = D_t(l, \nu)\eta^l, \quad -2 < l < \nu, \quad (7)$$

where

$$D_t(l, \nu) = \begin{cases} 0 & \nu > 2 \text{ \& } l \text{ is odd,} \\ \frac{\Gamma(\frac{l+2}{2})\Gamma(\frac{\nu-l}{2})}{\Gamma(\nu/2)} \nu^{\frac{l}{2}} & \text{otherwise.} \end{cases} \quad (8)$$

2.3. MGG Distribution

MGG process [19] can have different combinations, but we only consider the two-component case with the same shape parameter.

Suppose $q = \Re\{q\} + j\Im\{q\}$ following the zero-mean symmetric MGG process, its PDF is:

$$f(q) = (1 - \epsilon)f_1(q; \beta, \sigma) + \epsilon f_1(q; \beta, \tau\sigma), \quad (9)$$

where $f_1(q; \beta, \sigma) = \frac{\beta}{2\pi\sigma^2\Gamma(2/\beta)} \exp\left(-\frac{(\Re\{q\}^2 + \Im\{q\}^2)^{\frac{\beta}{2}}}{\sigma^\beta}\right)$. The $\epsilon \in (0, 1)$ is the weight parameter, $\beta > 0$ denotes the shape parameter tuning the decay rate of the density function, σ and $\tau\sigma$ with $\tau > 0$ are the scaling parameters of the two components. Note that when $\beta = 2$ or/and $\epsilon = 0, 1$, the MGG processes are reduced to Gaussian mixture (GM) and generalized Gaussian (GG) models, respectively.

The FLOMs of real and imaginary parts are:

$$E\{|q|^{l-2}\Re\{q\}^2\} = D_{\text{MGG}}(l, \beta, \epsilon, \tau)\sigma^l, \quad (10)$$

$$E\{|q|^{l-2}\Im\{q\}^2\} = D_{\text{MGG}}(l, \beta, \epsilon, \tau)\sigma^l, \quad (11)$$

where

$$D_{\text{MGG}}(l, \beta, \epsilon, \tau) = \begin{cases} 0 & \beta > 2 \text{ \& } l \text{ is odd,} \\ \frac{\Gamma(\frac{l+2}{\beta})(1-\epsilon+\epsilon\tau^l)}{\Gamma(2/\beta)} & \text{otherwise.} \end{cases} \quad (12)$$

3. Variance Analysis of Least ℓ_p -Norm Estimator

Without loss of generality, the complex-valued observed sequence $\mathbf{y} = [y_1 \cdots y_N]^T$ is modeled as

$$y_n = g_n(\mathbf{x}) + q_n, \quad (13)$$

where T is the transpose operator, $g_n(\mathbf{x}) = \Re\{g_n(\mathbf{x})\} + j\Im\{g_n(\mathbf{x})\}$ denotes an analytic function of \mathbf{x} with $\mathbf{x} = [x_1 \cdots x_M]^T$ being the deterministic parameter vector of interest, and q_n is the identically independent distributed (IID) noise component with zero location parameter. Assume that \mathbf{x} is real-valued, our task is to find \mathbf{x} from \mathbf{y} in presence of one of the non-Gaussian noise models, namely, SoS, Student's t and MGG.

To solve the problem, we concentrate on the complex-valued ℓ_p -norm with $p \geq 1$:

$$J(\mathbf{x}) = \sum_{n=1}^N |y_n - g_n(\mathbf{x})|^p = \sum_{n=1}^N \left((\Re\{y_n\} - \Re\{g_n(\mathbf{x})\})^2 + (\Im\{y_n\} - \Im\{g_n(\mathbf{x})\})^2 \right)^{\frac{p}{2}}. \quad (14)$$

3.1. Variance Formula Derivation

For the symmetric noise model, the ℓ_p -norm minimizer is always unbiased and the covariance matrix for $\hat{\mathbf{x}}$, denoted by $\mathbf{C}(\hat{\mathbf{x}})$, is [8]:

$$\mathbf{C}(\hat{\mathbf{x}}) = (E\{\mathbf{H}(J(\mathbf{x}))\})^{-1} E\{\nabla(J(\mathbf{x}))\nabla^H(J(\mathbf{x}))\} (E\{\mathbf{H}(J(\mathbf{x}))\})^{-1}, \quad (15)$$

where $\mathbf{H}(J(\mathbf{x}))$ and $\nabla(J(\mathbf{x}))$ are the Hessian matrix and gradient vector at the true value of \mathbf{x} , $^{-1}$ and H denote the matrix inverse and conjugate transpose operator, respectively. The variances of the estimates in $\hat{\mathbf{x}}$ are the diagonal elements of $\mathbf{C}(\hat{\mathbf{x}})$.

Utilizing (14), the expected values of the required derivatives in (15) are:

$$E\{\nabla(J(\mathbf{x}))\nabla^T(J(\mathbf{x}))\} = \frac{p^2}{2} \{ \nabla^T(\Re\{\mathbf{g}(\mathbf{x})\})E\{\Re\{\mathbf{V}\}\}\nabla(\Re\{\mathbf{g}(\mathbf{x})\}) + \nabla^T(\Im\{\mathbf{g}(\mathbf{x})\})E\{\Im\{\mathbf{V}\}\}\nabla(\Im\{\mathbf{g}(\mathbf{x})\}) \}, \quad (16)$$

$$E\{\mathbf{H}(J(\mathbf{x}))\} = \frac{p^2}{2} \{ \nabla^T(\Re\{\mathbf{g}(\mathbf{x})\})E\{\Re\{\mathbf{W}\}\}\nabla(\Re\{\mathbf{g}(\mathbf{x})\}) + \nabla^T(\Im\{\mathbf{g}(\mathbf{x})\})E\{\Im\{\mathbf{W}\}\}\nabla(\Im\{\mathbf{g}(\mathbf{x})\}) \}, \quad (17)$$

where $\Re\{\mathbf{V}\}$, $\Im\{\mathbf{V}\}$, $\Re\{\mathbf{W}\}$ and $\Im\{\mathbf{W}\}$ are $N \times N$ diagonal matrices with n -th elements $\Re\{V_n\} = E\{|q_n|^{\frac{p-2}{2}}\Re\{q_n\}^2\}$, $\Im\{V_n\} = E\{|q_n|^{\frac{p-2}{2}}\Im\{q_n\}^2\}$, $\Re\{W_n\} = E\{|q_n|^{\frac{p-4}{4}}\Re\{q_n\}^2\}$ and $\Im\{W_n\} = E\{|q_n|^{\frac{p-4}{4}}\Im\{q_n\}^2\}$, respectively.

Employing (15)–(17) as well as FLOM expressions in Section 2, the covariance matrix expression of the ℓ_p -norm minimizer in presence of three distributions, namely, SoS, Student's t and MGG processes, can be derived as:

$$\mathbf{C}_\alpha(\hat{\mathbf{x}}) = \frac{2\Gamma^2(2 - \frac{p}{2})\Gamma(p)\Gamma(1 - \frac{2p-2}{\alpha})\gamma^{\frac{2}{\alpha}}}{\Gamma^2(\frac{p}{2} + 1)\Gamma(2 - p)\Gamma^2(1 - \frac{p-2}{\alpha})} \mathbf{G}^{-1}, \quad 1 \leq p < \alpha, \quad (18)$$

$$\mathbf{C}_t(\hat{\mathbf{x}}) = \frac{\nu\Gamma(\frac{\nu}{2})\Gamma(p)\Gamma(\frac{\nu-2p+2}{2})\eta^2}{\Gamma^2(1 + \frac{p}{2})\Gamma^2(\frac{\nu-p+2}{2})} \mathbf{G}^{-1}, \quad p \geq 1 \quad (19)$$

$$\mathbf{C}_{\text{MGG}}(\hat{\mathbf{x}}) = \frac{\Gamma\left(\frac{2p}{\beta}\right)\Gamma\left(\frac{2}{\beta}\right)\left((1 - \epsilon) + \epsilon\tau^{2p-2}\right)\sigma^2}{\beta^2\Gamma^2\left(1 + \frac{p}{\beta}\right)\left((1 - \epsilon) + \epsilon\tau^{p-2}\right)^2} \mathbf{G}^{-1}, \quad p \geq 1. \quad (20)$$

where $\mathbf{G} = \nabla^T(\Re\{\mathbf{g}(\mathbf{x})\})\nabla(\Re\{\mathbf{g}(\mathbf{x})\}) + \nabla^T(\Im\{\mathbf{g}(\mathbf{x})\})\nabla(\Im\{\mathbf{g}(\mathbf{x})\})$ with $\Re\{\mathbf{g}(\mathbf{x})\} = [\Re\{g_1(\mathbf{x})\} \cdots \Re\{g_N(\mathbf{x})\}]^T$ and $\Im\{\mathbf{g}(\mathbf{x})\} = [\Im\{g_1(\mathbf{x})\} \cdots \Im\{g_N(\mathbf{x})\}]^T$. Notice that when ν and β are larger than 2, (19) and (20) are invalid for odd p .

In the following, two illustrations including the linear and nonlinear models are provided to elaborate the covariance expression.

Linear estimation

Here we consider the linear estimation problem with the observed signal:

$$y_n = A \exp(j(\omega n + \theta)) + q_n, \quad n = 1, 2, \dots, N, \quad (21)$$

where $A > 0$, $\omega \in (-\pi, \pi)$ and $\theta \in [0, 2\pi]$ denote the amplitude, frequency and phase, respectively. Assuming that ω is known, our task is to estimate A and θ .

Let $x = A \exp(j\theta)$. We easily get:

$$\mathbf{G} = N. \quad (22)$$

According to [20], the variances of \hat{A} and $\hat{\theta}$ are:

$$\text{var}(\hat{A}) \approx C(\hat{x}), \quad (23)$$

$$\text{var}(\hat{\theta}) \approx A^2 C(\hat{x}), \quad (24)$$

where $C(\hat{x})$ is calculated using (18)–(20) according to the respective noise model.

Nonlinear estimation

We now consider the nonlinear estimation problem. The data model is identical to (21). However, here the unknown parameters are ω and ϕ while A is known. In this case, we have $\mathbf{x} = [\omega \ \phi]^T$. Then we obtain [21]

$$\mathbf{G} = A^2 \begin{bmatrix} \frac{N(N+1)(2N+1)}{6} & \frac{N(N+1)}{2} \\ \frac{N(N+1)}{2} & \frac{N(N+1)(2N+1)}{6} \end{bmatrix}. \quad (25)$$

The corresponding covariance expressions for S α S, Student's t and MGG noise models are computed by (18)–(20), respectively.

3.2. Minimum Variance Analysis

We investigate the minimum achievable variance for each studied noise model using the ℓ_p -norm minimizer, via selection of optimum p . This is achieved by minimizing the scalar terms in (18)–(20) because the matrix components are independent of p .

3.2.1. S α S

We first investigate the ℓ_p -norm minimizer under S α S noise. For (18), the scalar term related to p , denoted by $H_\alpha(p)$, is:

$$H_\alpha(p) = \frac{\Gamma^2(2 - \frac{p}{2})\Gamma(p)\Gamma(1 - \frac{2p-2}{\alpha})}{\Gamma^2(\frac{p}{2} + 1)\Gamma(2 - p)\Gamma^2(1 - \frac{p-2}{\alpha})}. \quad (26)$$

To find the optimal value of p , denoted by p^* , we set the derivative of $\log(H_\alpha(p))$, denoted by $h_\alpha(p)$, to zero. Then $h_\alpha(p)$ has the form of

$$h_\alpha(p) = \psi(2 - p) - \psi\left(2 - \frac{p}{2}\right) + \psi(p) - \psi\left(\frac{p}{2}\right) + \frac{2}{\alpha} \left(\psi\left(1 - \frac{p-2}{\alpha}\right) - \psi\left(1 - \frac{2p-2}{\alpha}\right) \right) - \frac{2}{p} \quad (27)$$

where $\psi(\cdot)$ denotes the digamma function [22].

Let $u = p - 1 \in [0, 1)$. Employing the property that $\psi(x+1) = \psi(x) + \frac{1}{x}$, (27) is reduced to

$$h_\alpha(u) = \psi(1-u) - \psi\left(\frac{1-u}{2}\right) + \psi(1+u) - \psi\left(\frac{u+1}{2}\right) + \frac{2}{\alpha} \left(\psi\left(\frac{1-u}{\alpha}\right) - \psi\left(1 - \frac{2u}{\alpha}\right) \right) - \frac{2}{u+1}. \quad (28)$$

Utilizing $2\psi(2x) = 2\log(2) + \psi(x) + \psi(x + \frac{1}{2})$, we further simplify (28) as

$$h_\alpha(u) = h_1(u) + h_2(u) + h_3(u) + 4\log(2), \quad (29)$$

where

$$h_1(u) = -\psi(1-u) - \psi(u+1) + \psi\left(1 - \frac{u}{2}\right) + \psi\left(1 + \frac{u}{2}\right), \quad (30)$$

$$h_2(u) = \frac{2}{\alpha} \left(\psi\left(\frac{1-u}{\alpha}\right) - \psi\left(1 - \frac{2u}{\alpha}\right) \right), \quad (31)$$

$$h_3(u) = -\frac{2}{u+1}. \quad (32)$$

Taking the rational zeta series [21] on $h_1(u)$ yields

$$h_1(u) = \sum_{k=1}^{\infty} \zeta(k+1) \left\{ (-1)^k + 1 - \left(-\frac{1}{2}\right)^k - \left(\frac{1}{2}\right)^k \right\} u^k, \quad (33)$$

where $\zeta(\cdot)$ is the Riemann zeta function. Furthermore, based on the fact that $\psi(x) = -a + \sum_{k=0}^{\infty} \left(\frac{1}{k+1} - \frac{1}{k+x} \right)$ where a is the Euler-Mascheroni constant, as well as TSE at $u = 0$, $h_2(u)$ is rewritten as

$$h_2(u) = \frac{2}{\alpha} \left(\psi\left(\frac{1}{\alpha}\right) - \psi(1) \right) + \sum_{k=1}^{\infty} \frac{2u^k}{\alpha^{k+1}} \left\{ 2^k \zeta(k+1, 1) - \zeta\left(k+1, \frac{1}{\alpha}\right) \right\}. \quad (34)$$

where $\zeta(\cdot, \cdot)$ denotes the Hurwitz zeta function. Utilizing the TSE at $u = 0$, $h_3(u)$ is

$$h_3(u) = -2 - 2 \sum_{k=1}^{\infty} (-u)^k, \quad (35)$$

Combining (29)–(35), $h_\alpha(u)$ has the form of

$$h_\alpha(u) = b(0) + \sum_{k=1}^{\infty} b(k)u^k, \quad (36)$$

where $b(0) = 4\log(2) - 2 + \frac{2}{\alpha} (\psi(\frac{1}{\alpha}) - \psi(1))$ and

$$b(k) = \left((-1)^k + 1 - \left(\frac{1}{2}\right)^k - \left(-\frac{1}{2}\right)^k \right) \zeta(k+1) - 2(-1)^k + \left(\frac{1}{\alpha}\right)^k \left(2^k \zeta(k+1, 1) - \zeta\left(k+1, \frac{1}{\alpha}\right) \right). \quad (37)$$

According to extensive simulations, we find that the third-order polynomial in (36) can describe $h_\alpha(u)$ well. Therefore, we can obtain u^* by finding the real-valued root of the cubic function ($k = 3$) [23], which is

$$u^* = \sqrt[3]{-\frac{y}{2} + \sqrt{\left(\frac{y}{2}\right)^2 + \left(\frac{z}{3}\right)^3}} - \sqrt[3]{\frac{y}{2} + \sqrt{\left(\frac{y}{2}\right)^2 + \left(\frac{z}{3}\right)^3}}, \quad (38)$$

with $y = -2 \left(\frac{b(2)}{3b(3)} \right)^3 - \frac{b(1)b(2)}{3(b(3))^2} + \frac{b(0)}{b(3)}$ and $z = -\frac{1}{3} \left(\frac{b(2)}{b(3)} \right)^2 + \frac{b(1)}{b(3)}$.

Utilizing the definition of u , p^* is computed by $u^* + 1$. Note that for $\alpha \in [1, 2)$, the term $\left(\frac{y}{2}\right)^2 + \left(\frac{z}{3}\right)^3$ is larger than 0 and u^* is the only real root of (36). Since $h_\alpha(p)$ is monotonic, p^* is in the global optimum.

3.2.2. Student's t

From (19), the scalar term corresponding to Student's t noise, denoted by $H_t(p)$, is

$$H_t(p) = \frac{\Gamma(p)\Gamma(\frac{\nu-2p+2}{2})}{\Gamma^2(1+\frac{p}{2})\Gamma^2(\frac{\nu-p+2}{2})}. \quad (39)$$

In a similar manner, the derivative of $\log(H_t(p))$ can be expressed as

$$h_t(p) = \psi(p) - \psi\left(1 + \frac{p}{2}\right) - \psi\left(\frac{\nu-2p}{2} + 1\right) + \psi\left(\frac{\nu-p}{2} + 1\right). \quad (40)$$

Here we consider two cases: $\nu < 2$ and $\nu \geq 2$. In the scenario of $\nu < 2$, we have

$$h'_t(p) = \psi^{(1)}(p) - \frac{1}{2}\psi^{(1)}\left(1 + \frac{p}{2}\right) + \psi^{(1)}\left(\frac{\nu-2p}{2} + 1\right) - \frac{1}{2}\psi^{(1)}\left(\frac{\nu-p}{2} + 1\right). \quad (41)$$

where $\psi^{(1)}(\cdot)$ is the trigamma function. Based on the multiplication theorem of trigamma function, (41) is rewritten as

$$\begin{aligned} h'_t(p) = & \frac{1}{4} \left(\psi^{(1)}\left(\frac{p}{2}\right) + \psi^{(1)}\left(\frac{1}{2} + \frac{p}{2}\right) \right) - \frac{1}{2}\psi^{(1)}\left(1 + \frac{p}{2}\right) - \frac{1}{2}\psi^{(1)}\left(\frac{\nu}{2} + 1 - \frac{p}{2}\right) \\ & + \frac{1}{4} \left(\psi^{(1)}\left(\frac{\nu}{4} + \frac{1}{2} - \frac{p}{2}\right) + \psi^{(1)}\left(\frac{\nu}{4} + 1 - \frac{p}{2}\right) \right) \end{aligned} \quad (42)$$

which is greater than 0. Therefore, we can deduce that $h_t(p) > h_t(1) > 0$. In this case, the optimum p is 1.

In the case of $\nu \geq 2$, assigning $u = p - 1 \in [0, 1)$, (40) is

$$h_t(u) = \psi(u+1) - \psi\left(1 + \frac{u+1}{2}\right) - \psi\left(\frac{\nu-2u}{2}\right) + \psi\left(\frac{\nu-u-1}{2} + 1\right). \quad (43)$$

Similarly to the SaS distribution, (43) can be written as the same expression as in (36) except that

$$b(0) = 2\log(2) - 2 - \frac{2}{\nu}, \quad (44)$$

$$b(k) = \zeta(k+1) \left((-1)^k - \left(-\frac{1}{2}\right)^k \right) - 2(-1)^k + \left\{ \zeta\left(k, \frac{\nu}{2}\right) \left(\left(\frac{1}{2}\right)^k - 1 \right) - \frac{2}{\nu^{k+1}} \right\}. \quad (45)$$

It can be also easily shown that the series in Student's t process converges when $k \geq 3$. The optimum value is $p^* = u^* + 1$ with u^* being calculated by (38).

3.2.3. MGG

For MGG noise, the corresponding scalar term, which is extracted from (20), has the form of

$$H_{\text{MGG}}(p) = \frac{\Gamma\left(\frac{2p}{\beta}\right) \left((1-\epsilon) + \epsilon\tau^{2p-2} \right)}{\Gamma^2\left(1 + \frac{p}{\beta}\right) \left((1-\epsilon) + \epsilon\tau^{p-2} \right)^2}. \quad (46)$$

Let $u = \frac{p}{\beta} - \frac{1}{2} \in (-1/2, 1/2)$, the derivative of $\log(H_{\text{MGG}}(p))$ is factorized as:

$$h_{\text{MGG}}(u) = \frac{1}{\beta} \left\{ \psi(2u+1) - \psi\left(\frac{3}{2} + u\right) \right\} + \ln(\tau) \left\{ \frac{1}{1 + \frac{\epsilon}{1-\epsilon}\tau^{\beta u + \beta/2 - 2}} - \frac{1}{1 + \frac{\epsilon}{1-\epsilon}\tau^{2\beta u + \beta - 2}} \right\}. \quad (47)$$

To solve $h_{\text{MGG}}(u) = 0$, we express (47) as (36), where $b(0)$ and $b(k)$ are now

$$b(0) = \frac{2\log(2) - 2}{\beta} \quad (48)$$

$$b(k) = \frac{1}{\beta} \left\{ \zeta(k+1) \left((-2)^k - (-1)^k \right) + (-2)^{k+1} \right\} - \ln(\tau) \left\{ I^{(k)}\left(\frac{\epsilon\tau^{\beta/2-2}}{(1-\epsilon)}, \tau^\beta\right) - I^{(k)}\left(\frac{\epsilon\tau^{\beta-2}}{(1-\epsilon)}, \tau^{2\beta}\right) \right\} \quad (49)$$

where $I^{(k)}(c, x)$ denotes k th-order derivative of $\frac{1}{1+cx^u}$ with respect to u when $u = 0$.

According to extensive simulations, the series here approximates $h_{\text{MGG}}(u)$ well when $k \geq 3$. Therefore, we derive the optimum p^* as $p^* = \beta(u^* + 1/2)$ with u^* having the same form in (38).

4. Simulation Results

To verify the derived variance formulas in the case of three symmetric impulsive noise distributions, computer simulations have been conducted. The signal is generated according to (21) and the corresponding parameters are $A = 1$, $\omega = 1.25$ and $\theta = 0.5$. For S α S, Student's t and MGG models, the corresponding density parameters are chosen as $\alpha = 1.6$ [24], $\nu = 5$ [25], and $\beta = 5$, $\epsilon = 0.01$, $\tau = \sqrt{5}$ [11]. The special case of MGG distribution, referred to as GM process, is also shown here with $\beta = 2$, $\epsilon = 0.01$ and $\tau = \sqrt{5}$. The GG model is not considered here because the optimal ℓ_p -norm minimizer is the ML estimator [14]. According to (38), the optimal value p^* for these four settings, are 1.16, 1.31, 2.75 and 1.86, respectively. To provide a comparison, the empirical optimum results in [14], referring to as $p = (1 + \alpha)/2$, $p = (\nu + 4)/5$, and $p = 1 + \frac{2}{100\epsilon+3}$, are investigated here. Furthermore, comparisons with the least ℓ_1 -norm estimator and the CRLB are also included. Since the second-order power diverges for the S α S model, we utilize the geometric signal-to-noise ratio (GSNR) to produce different noise conditions [26]. All results are based on 5000 Monte Carlo simulation trials with a data length of $N = 50$.

First of all, to investigate the correctness of cubic polynomial, we study the value of optimum p obtained by (36) and (38). According to [14], we choose degree 100 replacing the infinity order in (36). Figure 1 shows the comparison of p^* calculation between the cubic function and polynomial (36) with degree 100 versus α , ν , β , respectively. Other density parameters align with the previous set. It is experimentally observed in S α S, Student's t and MGG models that results by cubic function attain the true values well, which corroborates our claim in Section 3.

Secondly, we address linear estimation, namely, studying the amplitude and phase estimation performance using the complex-valued observations in (21). Figure 2 and 3 show MSE performance of amplitude and phase versus SNR/GSNR, respectively. In these figures, we see the near optimality of the least ℓ_p -norm estimator in the Student's t and GM noise models and its suboptimal performance in the S α S and MGG disturbance. Most importantly, in Figure 2 and 3, the gap between the optimal value and result in [14] indicates that our method is more accurate. Note that the gap between our proposal and [14] becomes bigger as β increases, verifying the importance of our method.

Thirdly, nonlinear estimation of exponential signal frequency and phase is studied. The MSE results are plotted in Figures 4 and 5. We again see that the variance formulas are validated and the other findings are similar to those of Figures 2 and 3. Furthermore, when SNR/GSNR ≥ 5 dB, our proposals perform better than the ℓ_1 -norm minimizer and results in [14].

In summary, in the presence of S α S, Student's t and MGG noises, our proposed method is superior to that in [14]. It has also been discussed in [14] that in the real applications that the density parameters of noise are unknown, the least ℓ_1 -norm estimator is a good choice. However, according to the simulations on the MGG distribution, the gap between the $p = 1$ and p^* is ≥ 3 dB, indicating the inferiority of the ℓ_1 -norm minimizer. In this case, the shape parameter should be estimated firstly.

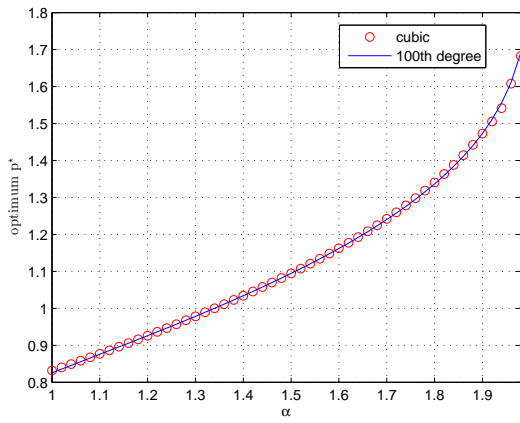
5. Conclusion

In this work, we focus on the variance expression deviation of the least ℓ_p -norm technique with $p \geq 1$, where three representative bivariate symmetric disturbances has been studied. The optimal choice of p , corresponding to the minimum variance, is discussed, which can be obtained by solving a specific cubic equation. It is worth pointing out that for the bivariate Student's t process with $\nu < 2$, the optimum p should be chosen as 1. Simulation results validate the accuracy of the derived variance formulas using linear and nonlinear estimation examples with complex-valued observations. It is also demonstrated that variances of the least ℓ_p -norm estimator for S α S and Student's t noise models are very close to corresponding CRLB. This result indicates that this estimator can provide optimum or nearly-optimum performance for these three noise models if an appropriate value of p is chosen. Note that all results in this paper can be applied to the real-valued scenario.

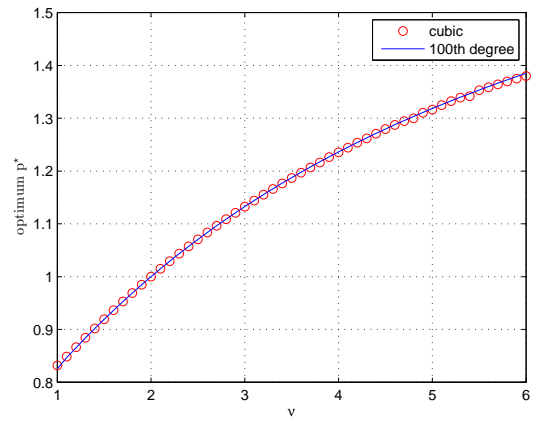
- [1] P.J. Schreier and L.L. Scharf, *Statistical Signal Processing of Complex-Valued Data: The Theory of Improper and Noncircular Signals*, Cambridge, U.K.: Cambridge University Press, 2010.
- [2] V. Madisetti, *Wireless, Networking, Radar, Sensor Array Processing, and Nonlinear Signal Processing*, Chapman & Hall/CRC Press, 2009.
- [3] E. Ollila, V. Koivunen and H.V. Poor, "Complex-valued signal processing-essential models, tools and statistics," in *Proceedings of 2011 Information Theory and Applications Workshop (ITA)*, La Jolla, CA, pp. 1-10, Feb. 2011.
- [4] S.M. Kay, *Fundamentals of Statistical Signal Processing: Estimation Theory*, Prentice-Hall, NJ: Englewood Cliffs, 1993.
- [5] P. Stoica and R. Moses, *Introduction of Spectral Analysis*, Upper Saddle River, NJ: Prentice-Hall, 1997.
- [6] A. Papoulis, *Probability, Random Variables and Stochastic Processes*, McGraw-Hill, 1991.
- [7] V.H. MacDonald and P.M. Schultheiss, "Optimum passive bearing estimation in a spatially incoherent noise environment," *Journal of the Acoustical Society of America*, vol. 46, no. 1, pt. 1, pp. 37-43, Jul. 1969.
- [8] H.C. So, Y.T. Chan, K.C. Ho and Y. Chen, "Simple formulas for bias and mean square error computation," *IEEE Signal Processing Magazine*, vol. 30, no. 4, pp. 162-165, Jul. 2013.
- [9] C.L. Nikias and M. Shao, *Signal Processing with Alpha-Stable Distribution and Applications*, New York: John Wiley & Sons Inc., 1995.
- [10] K. Samuel and N. Saralees, *Multivariate t -Distributions and their Applications*, Cambridge University Press, 2004.
- [11] M.O.M. Mahmoud, M. Jaidane-Saidane and N. Hizaoui, "The use of mixture of generalized Gaussian for trend analysis of the load duration curve: Summer and winter load variability in Tunisia," in *Proceedings of 43rd Universities Power Engineering Conference*, Padova, Italy, Sept. 2008, pp. 1-5.

- 1 [12] P.J. Huber and E.M. Ronchetti, *Robust Statistics*, Hoboken, N.J.: Wiley, 2009.
- 2
- 3 [13] T.H. Li, "A nonlinear method for robust spectral analysis," *IEEE Transactions on Signal Processing*, vol.
- 4 58, no. 5, pp. 2466-2474, May 2010.
- 5
- 6 [14] Y. Chen, H.C. So and E.E. Kuruoglu, "Variance analysis of unbiased least ℓ_p -norm estimator in non-
- 7 Gaussian noise," *Signal Processing*, vol. 122, pp. 190-203, 2016.
- 8
- 9 [15] A.H. Money, J.F. Affleck-Graves, M.L. Hart and G.D.I. Barr, "The linear regression model: ℓ_p -norm
- 10 estimation and the choice of p ," *Communications in Statistics-Simulation and Computation*, vol. 11, no.
- 11 1, pp. 89-109, 1982.
- 12
- 13 [16] E.E. Kuruoglu, P.J.W. Rayner and W.J. Fitzgerald, "Least ℓ_p -norm impulsive noise cancellation with
- 14 polynomial filters," *Signal Processing*, vol. 69, no. 1, pp. 1-14, Aug. 1998.
- 15
- 16 [17] E. Ollila, D. E. Tyler, V. Koivunen and H. V. Poor, "Complex elliptically symmetric distributions: Survey,
- 17 new results and applications," *IEEE Transactions on Signal Processing*, vol. 60, no. 11, pp. 5597-5625, Nov.
- 18 2012.
- 19
- 20 [18] X.Y. Ma and C.L. Nikias, "Parameter estimation and blind channel identification in impulsive signal
- 21 environments," *IEEE Transactions on Signal Processing*, vol. 43, no. 12, pp. 2884-2897, Dec. 1995.
- 22
- 23 [19] M.S. Allili, "Wavelet modeling using finite mixtures of generalized Gaussian distributions: Application to
- 24 texture discrimination and retrieval," *IEEE Transactions on Image Processing*, vol. 21, no. 4, pp. 1452-1464,
- 25 Apr. 2012.
- 26
- 27 [20] E. Anarim and Y. I Stefanopulos, "Statistical analysis of Pisarenko type tone frequency estimator," *Signal*
- 28 *Processing*, vol. 24, no. 1, pp. 291-298, 1991.
- 29
- 30 [21] I.D. Gradshteyn and I.M. Ryzhik, *Tables of Integrals, Series and Products*, New York: Academic, 2007.
- 31
- 32 [22] M. Abramowitz and I.A. Stegun, *Handbook of Mathematical Functions*, New York: Dover Publications,
- 33 1964.
- 34
- 35 [23] J. Nathan, *Basic Algebra I*, New York: Dover Publications, 2009.
- 36
- 37 [24] P. G. Georgiou, P. Tsakalides and C. Kyriakakis, "Alpha-stable modeling of noise and robust time delay
- 38 estimation in the presence of impulsive noise," *IEEE Transactions on Multimedia*, vol. 1, no. 3, pp. 291-301,
- 39 Sep. 1999.
- 40
- 41 [25] G.J. Lauprete, A.M. Samarov and R. E. Welsch, "Robust portfolio optimization," *Metrika*, vol. 55, no. 1,
- 42 pp. 139-149, Apr. 2002.
- 43
- 44 [26] J.G. Gonzalez, J.L. Paredes and G.R. Arce, "Zero-order statistics: A mathematical framework for the
- 45 processing and characterization of very impulsive signals," *IEEE Transactions on Signal Processing*, vol.
- 46 54, no. 10, pp. 3839-3851, Oct. 2006.
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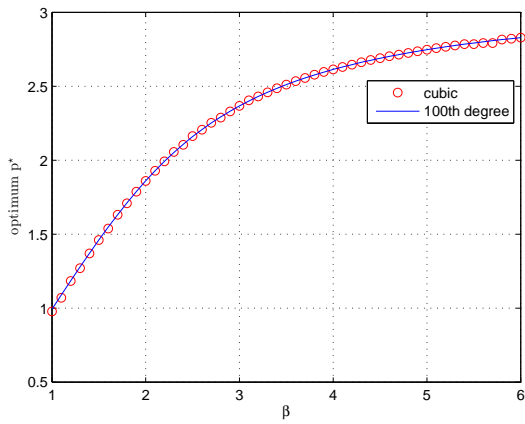
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a. SaS



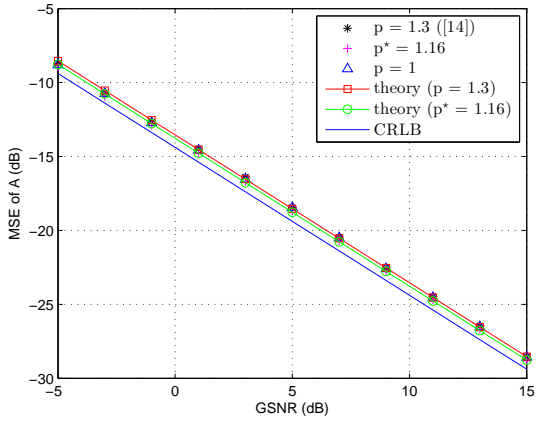
b. Student's t



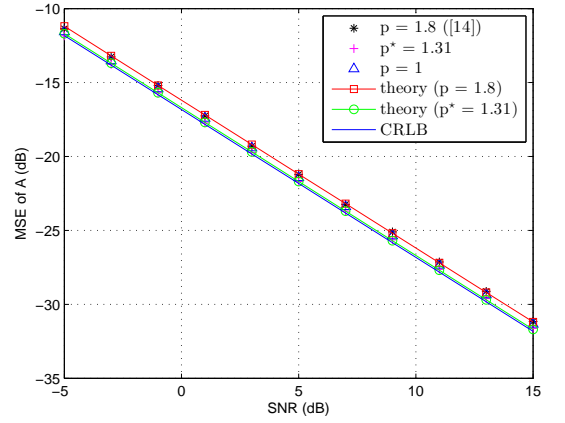
c. MGG

Figure 1: Optimum p^* calculated by cubic and 100-th degree polynomials versus density parameters

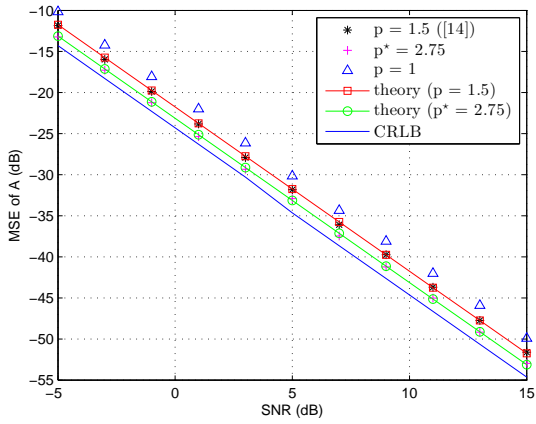
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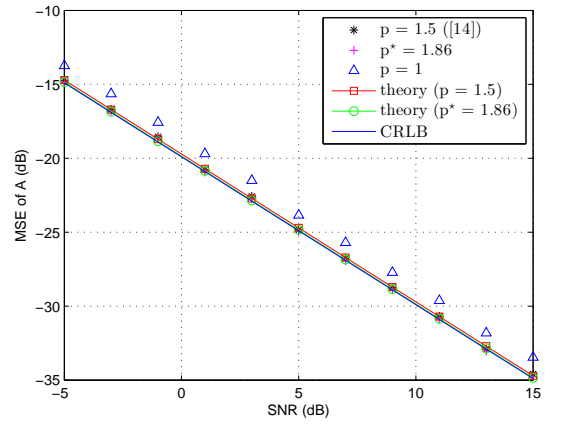
a. S α S



b. Student's t



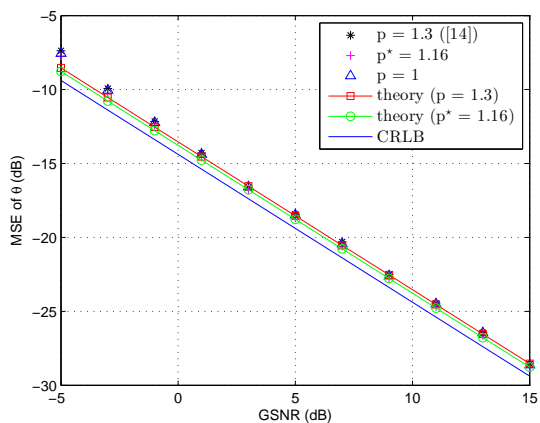
c. MGG



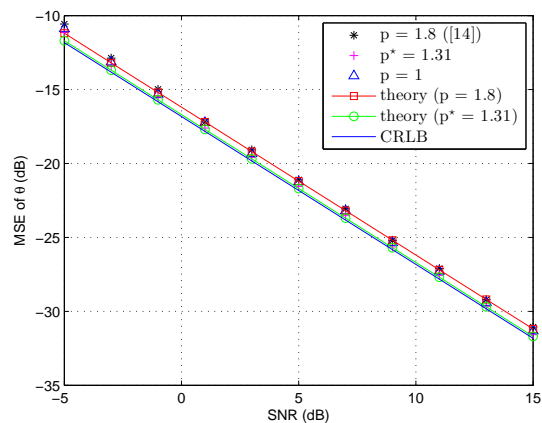
d. GM

Figure 2: MSE of sinusoidal amplitude in complex-valued observations

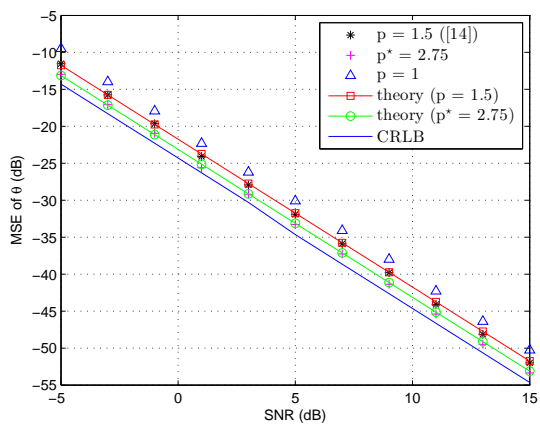
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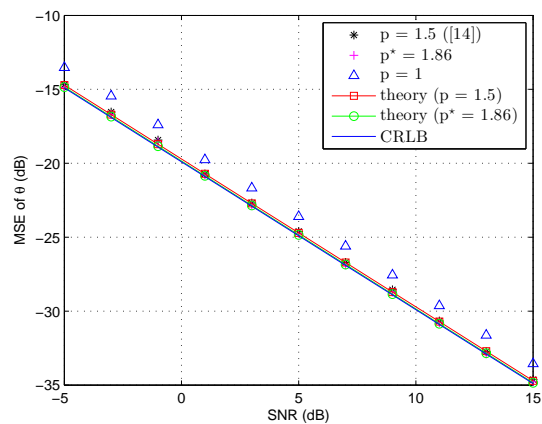
a. SaS



b. Student's t



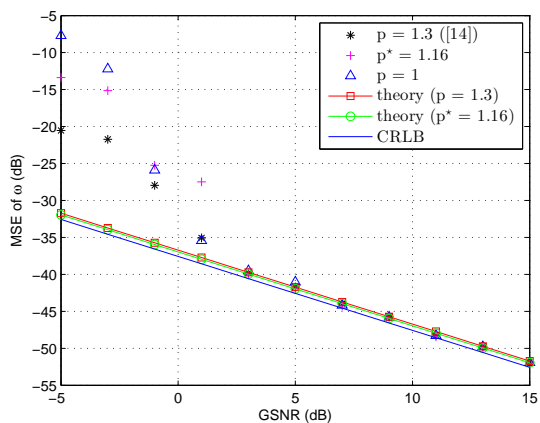
c. MGG



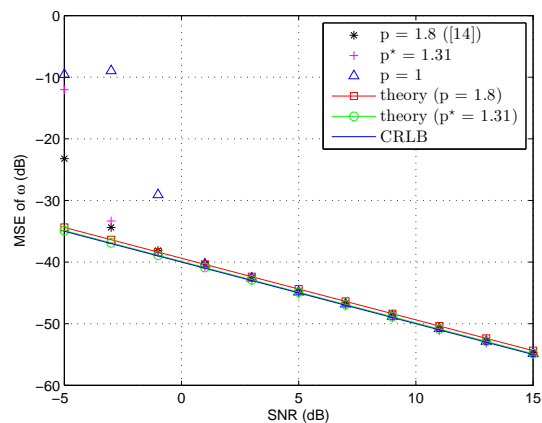
d. GM

Figure 3: MSE of sinusoidal phase in complex-valued observations

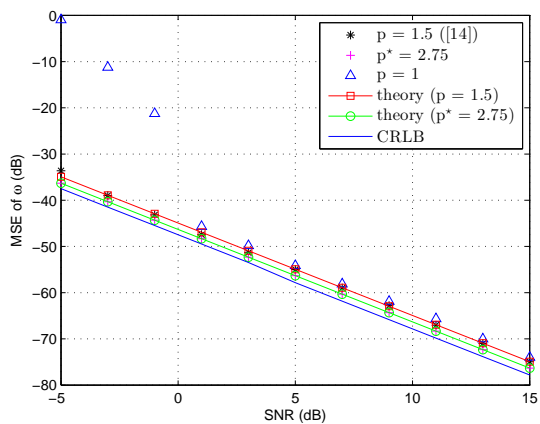
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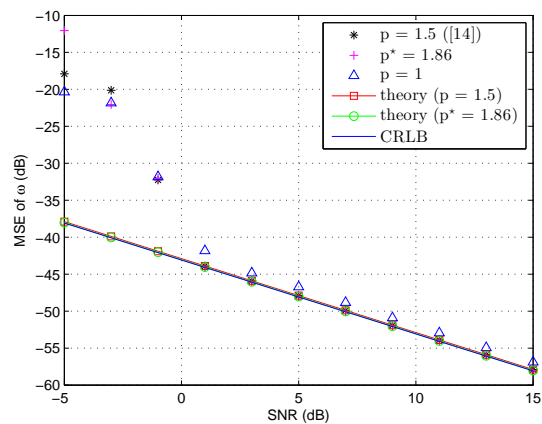
a. SaS



b. Student's t



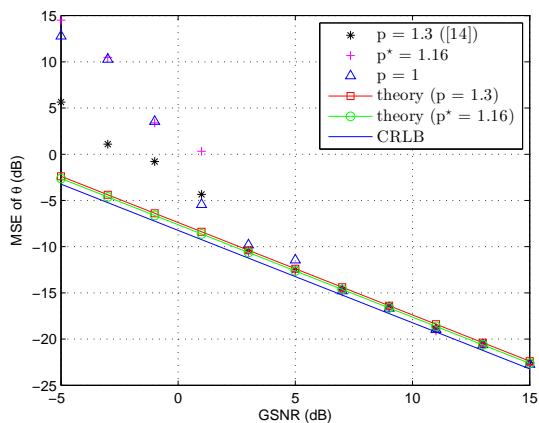
c. MGG



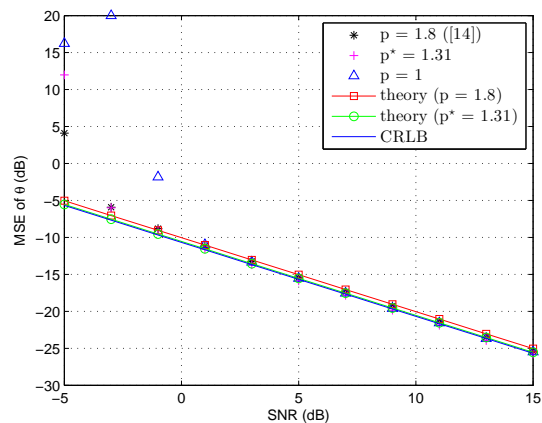
d. GM

Figure 4: MSE of sinusoidal frequency in complex-valued observations

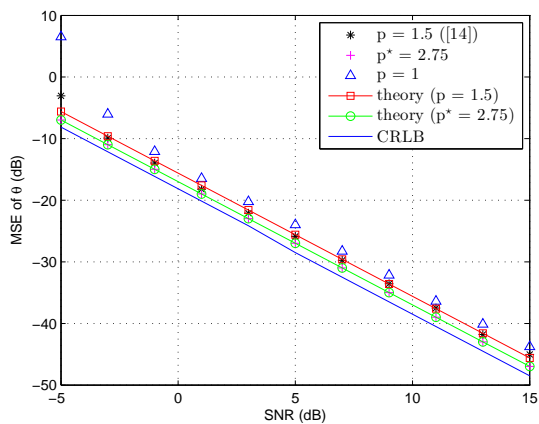
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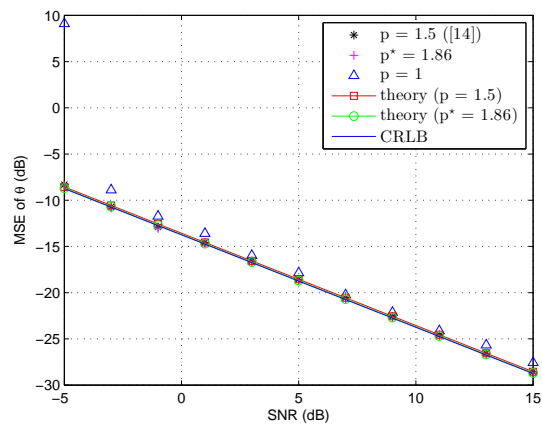
a. SaS



b. Student's t



c. MGG



d. GM

Figure 5: MSE of sinusoidal phase in complex-valued observations